Time-integration methods for finite element discretisations of the second-order Maxwell equation

To the memory of Jan Verwer

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Abstract

This article deals with time integration for the second-order Maxwell equations with possibly non-zero conductivity in the context of the discontinuous Galerkin finite element method (DG-FEM) and the $H$(curl)-conforming FEM. For the spatial discretisation, hierarchical $H$(curl)-conforming basis functions are used up to polynomial order $p = 3$ over tetrahedral meshes, meaning fourth-order convergence rate. A high-order polynomial basis often warrants the use of high-order time-integration schemes, but many well-known high-order schemes may suffer from a severe time-step stability restriction owing to the conductivity term. We investigate several possible time-integration methods from the point of view of accuracy, stability and computational work. We also carry out a numerical Fourier analysis to study the dispersion and dissipation properties of the semi-discrete DG-FEM scheme as well as the fully-discrete schemes with several of the time-integration methods. The dispersion and dissipation properties of the spatial discretisation and those of the time-integration methods are investigated separately, providing additional insight into the two discretisation steps.

Keywords: $H$(curl)-conforming finite element method, discontinuous Galerkin finite element method, high-order numerical time integration, second-order damped Maxwell wave equation

1. Introduction

High-order finite element methods (FEM) are an increasingly important technology in large-scale electromagnetic simulations thanks to their ability to effectively model complex geometrical structures and long-time wave propagation. It has long been known that the standard $H^1$-conforming FEM for electromagnetic waves may result in non-physical, spurious solutions. Instead, one may naturally opt for the $H$(curl)-conforming FEM pioneered by Nédélec [1, 2] and Bossavit [3, 4]. It has the advantage of mimicking the

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Figure 1: Sparsity pattern of the mass matrix for \( H(\text{curl}) \)-conforming FEM (left) and DG-FEM (right) for a mesh with 320 elements. Third-order polynomials are used, which means that the size of the blocks in the right plot is \( 60 \times 60 \). Note the difference in size between the two matrices.

Geometrical properties of the Maxwell equations at the discrete level. However, in time-domain computations it requires solving linear systems with mass matrices even if explicit time integration is employed. One attractive alternative – also free of spurious solutions under certain conditions – is the discontinuous Galerkin FEM (DG-FEM) \([5, 6, 7]\), where the resulting mass matrix is block-diagonal and therefore the computational cost of its inversion is negligible. This additional flexibility, however, comes at a cost. The number of degrees of freedom in DG discretisations is higher than that in the \( H(\text{curl}) \)-conforming discretisation, although the difference decreases with the polynomial order in the spatial discretisation. As an illustration, Figure 1 shows the sparsity patterns of the mass matrices for both methods when a mesh of 320 tetrahedra and third-order polynomials are used. So there appears to be a trade-off between the two methods in time-domain computations. In general, the \( H(\text{curl}) \)-conforming approach is more efficient with low-order polynomials and DG-FEM with high-order ones \([8, 9]\). The expected break-even point depends on a number of factors, such as the conditioning and sparsity of the mass and stiffness matrices in the resulting semi-discrete systems.

Throughout the article, the different discretisation techniques are applied to the three-dimensional Maxwell equations in the second-order time-dependent form,

\[
\varepsilon_r \frac{\partial^2 \mathbf{E}}{\partial t^2} + \sigma \frac{\partial \mathbf{E}}{\partial t} + \nabla \times \left( \mu_r^{-1} \nabla \times \mathbf{E} \right) = -\frac{\partial \mathbf{J}}{\partial t},
\]

with perfectly electric conducting boundary conditions \( \mathbf{n} \times \mathbf{E} = 0 \). All quantities are dimensionless in (1), where \( \mathbf{E} \) is the electric field and \( \mathbf{J} \) is the electric current density.\(^1\)

\(^1\)We can derive the dimensionless form by using the scalings \( x = \tilde{x}/L, t = \tilde{t}/(L/\tilde{c}_0), \mathbf{E} = \mathbf{E}/(\tilde{Z}_0 H_0), \mathbf{H} = \mathbf{H}/H_0, \mathbf{J} = \mathbf{J}/(H_0/L) \) and \( \sigma = J L \tilde{c}_0 /E \), with tilde denoting the dimensional quantities. Here \( \tilde{L} \) is the reference length, \( \tilde{c}_0 = (\mu_0 \tilde{\mu}_0)^{-1/2} \) is the speed of light in vacuum, \( \mathbf{H} \) is the magnetic field
The values $\sigma$, $\varepsilon_r$ and $\mu_r$ are assumed to be time-independent constant scalars, and they respectively denote conductivity, relative dielectric permittivity and relative magnetic permeability. If the domain is filled with non-conductive material, the damping term $\sigma \frac{\partial \mathbf{E}}{\partial t}$ is absent. If, in addition, the source term $-\partial_t \mathbf{J}$ is also taken to be zero, we have the conservative Maxwell wave equation.

Following the method of lines, we first discretise the spatial operators with the $H(\text{curl})$-conforming FEM or the DG-FEM. In either case, we arrive at a semi-discrete system in the form of second-order ordinary differential equations (ODEs) in $\mathbb{R}^n$,

$$M_e u'' + M_\sigma u' + S_\mu u = j,$$

where $u$ is the unknown vector of $N$ scalar coefficients associated with the approximation of the electric field $\mathbf{E}$. The source term $j$ is the projection of $-\partial_t \mathbf{J}$ onto the finite-element space and in general may also contain boundary data. For simplicity, however, we restrict ourselves to the homogeneous Dirichlet boundary condition, $\mathbf{n} \times \mathbf{E} = 0$, in this article. Each term in the left-hand side of (2) corresponds to the respective term in the left-hand side of (1). The mass matrix $M_e$ is symmetric positive definite and the conductivity matrix $M_\sigma$ is symmetric positive semi-definite. In addition, for constant scalars $\sigma$ and $\varepsilon_r$, the matrices $M_e$ and $M_\sigma$ are identical up to a constant. The stiffness matrix $S_\mu$ is the discretisation of the wave term and is symmetric positive semi-definite.

A key feature of (1) and (2) from the point of view of time integration is that it includes the conductivity $\sigma$. Even moderate values of $\sigma$ may result in a prohibitively small time step for many of the popular time-integration schemes. Therefore, we pay special attention to time-integration methods that treat the conductivity mass matrix $M_\sigma$ in an implicit way. Many of such methods and others discussed in this article have been previously studied in [10] for the system of first-order Maxwell equations discretised by the lowest-order $H(\text{curl})$-conforming elements. See also [11] for more details on high-order composition methods for the conduction-free Maxwell equations.

The semi-discrete system (2) conserves (discrete) energy for the spatial discretisations discussed here, since these are both symmetric in the matrix/operator sense. Hence, using an energy-conservative time-integration method results in a conservative fully-discrete scheme. We investigate the dispersion and dissipation error of the schemes in two steps. First, we determine the dispersion error of the semi-discrete scheme by solving the time-harmonic eigenvalue problem corresponding to the semi-discrete system. Second, we can then apply any given time-integration scheme to a simple, but equivalent, model problem that includes the information of the semi-discrete numerical frequency, and thus define the dispersion (and, if there is any, dissipation) error of the time-integration method. This approach shows whether the dispersion error is dominated by the spatial or temporal discretisation — a piece of information that may prove useful in deciding whether or not to go for high-order time-integration schemes.

The computational performance of the $H(\text{curl})$-conforming method hinges to a great degree on efficiently solving the linear system with the mass matrix. A number of advanced

(eliminated in (1)), $\mathbf{H}_0$ is the reference magnetic field strength and $\tilde{Z}_0 = (\tilde{\omega} \mu_0 / (\tilde{\sigma} + \tilde{\omega} \varepsilon_0))^{1/2}$ is the intrinsic impedance, with $\tilde{\omega}$ being the angular frequency and $i$ the imaginary unit.
techniques have been proposed recently, including special mass lumping \[12, 13, 14\], the explicit computation of an approximate sparse inverse mass matrix \[15\], or the construction of special preconditioners. These approaches, however, do not in their current form seem to provide a general framework and therefore cannot always be extended to high-order discretisations in a straightforward manner. For this reason in this article we resort to standard preconditioners. It is, of course, also possible to use sparse direct solvers but they are still very often found to be too memory extensive for large-scale three-dimensional computations.

The remaining part of the article is organised as follows. The weak formulations of the \(H(\text{curl})\)-conforming FEM and the DG-FEM are given in Section 2. The semi-discrete system arising from either of the spatial discretisations is analysed in Section 3, while we briefly describe a number of the most widely-used time-integration methods in Section 4. Numerical examples that compare the computational performance of the two finite element approaches are presented in Section 5, while Section 6 investigates the numerical dispersion and dissipation properties of the semi-discrete as well as the fully discrete schemes. Section 7 concludes the article with final remarks.

2. The weak formulation

To present the weak formulations that result from the \(H(\text{curl})\)-conforming and the DG discretisations, we introduce the tessellation \(\mathcal{T}_h\) that partitions the polyhedral domain \(\Omega \subset \mathbb{R}^3\) into a set of tetrahedra \(\{K\}\). Throughout the article we assume that the mesh is shape-regular and that each tetrahedron is straight-sided. The notations \(\mathcal{F}_h, \mathcal{F}_h^i\) and \(\mathcal{F}_h^b\) stand respectively for the set of all faces \(\{F\}\), the set of all internal faces, and the set of all boundary faces.

On the computational domain \(\Omega\), we define the spaces

\[
H(\text{curl}; \Omega) := \left\{ u \in \left[ L^2(\Omega) \right]^3 : \nabla \times u \in \left[ L^2(\Omega) \right]^3 \right\},
\]

\[
H_0(\text{curl}; \Omega) := \left\{ u \in H(\text{curl}; \Omega) \middle| n \times u = 0 \text{ on } \partial \Omega \right\},
\]

and the \(L^2\) inner product \((\cdot, \cdot)\)

\[
(u, v) = \int_{\Omega} u \cdot v \, dV.
\] (3)

The continuous weak formulation of (1) now reads as follows: Find \(E \in H_0(\text{curl}, \Omega)\) such that \(\forall w \in H_0(\text{curl}, \Omega)\) the relation

\[
\frac{\partial^2}{\partial t^2} (\varepsilon_t E, w) + \frac{\partial}{\partial t} (\sigma E, w) + (\mu^{-1} \nabla \times E, \nabla \times w) = - \left( \frac{\partial J}{\partial t} , w \right)
\] (4)

is satisfied. See e.g. \[16, 17\].

2.1. Weak formulation of the globally \(H(\text{curl})\)-conforming discretisation

In order to discretise (4), we first introduce the finite element space associated with the tessellation \(\mathcal{T}_h\). Let \(\mathcal{P}_p(K)\) be the space of polynomials of degree at most \(p \geq 1\) on
\( K \in T_h \). Over each element \( K \) the \( H(\text{curl}) \)-conforming polynomial space is defined as

\[
Q^p = \left\{ \mathbf{u} \in [P_p(K)]^3; \; \mathbf{u}_T|_{F^K} \in [P_p(F^K_i)]^2; \; \mathbf{u} \cdot \mathbf{t}_j|_{e^K_j} \in P_p(e^K_j) \right\},
\]

where \( F^K_i \), \( i = 1, 2, 3, 4 \) are the faces of the element; \( e^K_j \), \( j = 1, 2, 3, 4, 5, 6 \) are the edges of the element; \( \mathbf{u}_T \) is the tangential component of \( \mathbf{u} \); and \( \mathbf{t}_j \) is the directed tangential vector on edge \( e^K_j \). For the construction of \( Q^p \), we use a set of \( H(\text{curl}) \)-conforming hierarchic basis functions [18, 19].

Next, we introduce the discrete space of globally \( H(\text{curl}) \)-conforming functions

\[
\mathcal{Y}^p_h := \left\{ \mathbf{u} \in [H_0(\text{curl}, \Omega)]^3 \; \big|\; \mathbf{u}|_K \in Q^p, \; \forall K \in T_h \right\},
\]

and let the set of basis functions \( \{ \psi_i \} \) span the space \( \mathcal{Y}^p_h \). See [17] for a detailed discussion on both continuous and discrete \( H(\text{curl}) \)-conforming spaces. We can then approximate the electric field \( \mathbf{E} \) as

\[
\mathbf{E} \approx \mathbf{E}_h = \sum_i u_i(t) \psi_i(x),
\]

from which the discrete weak formulation reads as follows: Find \( \mathbf{E}_h \in \mathcal{Y}^p_h \) such that \( \forall \phi \in \mathcal{Y}^p_h \) the relation

\[
\frac{\partial^2}{\partial t^2} (\varepsilon \mathbf{E}_h, \phi) + \frac{\partial}{\partial t} (\sigma \mathbf{E}_h, \phi) + (\mu^{-1} \nabla \times \mathbf{E}_h, \nabla \times \phi) = - \left( \frac{\partial \mathbf{J}}{\partial t}, \phi \right)
\]

is satisfied. Note that (7) is satisfied if and only if it is satisfied for every basis function \( \psi_i \), \( i = 1, \ldots, N \), with \( N \) being the global number of degrees of freedom. As a result, substitution of (6) into (7) yields the semi-discrete system (2) with

\[
[M_e]_{ij} = (\varepsilon \psi_i, \psi_j), \quad [S_\mu]_{ij} = (\mu^{-1} \nabla \times \psi_i, \nabla \times \psi_j), \quad [M_\sigma]_{ij} = (\sigma \psi_i, \psi_j), \quad [j]_i = - \left( \frac{\partial \mathbf{J}}{\partial t}, \psi_i \right).
\]

Each of the above matrices \( - M_e, M_\sigma \) and \( S_\mu \) has a large number of entries far off the diagonal, increasing the computational cost for both explicit and implicit time-integration methods.

2.2. Weak formulation of DG-FEM

In contrast to the \( H(\text{curl}) \)-conforming discretisation, in DG-FEM we are looking for the discrete solution in the space

\[
\Sigma^p_h := \left\{ \sigma \in [L^2(\Omega)]^3 \; \big|\; \sigma|_K \in Q^p, \; \forall K \in T_h \right\}
\]

with \( Q^p \) defined in (5). That is, we allow the polynomial functions to be fully discontinuous across element interfaces and assume that the set of basis functions \( \{ \psi_i \} \) now span the space \( \Sigma^p_h \). Instead of enforcing continuity of the tangential components, the information between elements is now coupled through the numerical flux [5, 20, 7]. Before we can
define the numerical flux and formulate the discretisation for DG-FEM, we first need to introduce more notation.

Consider an interface \( F \in \mathcal{F}_h \) between element \( K^L \) and element \( K^R \), and let \( \mathbf{n}^L \) and \( \mathbf{n}^R \) represent their respective outward pointing normal vectors. We define the tangential jump and the average of the quantity \( \mathbf{u} \) across interface \( F \) as

\[
[u]_T = \mathbf{n}^L \times \mathbf{u}^L + \mathbf{n}^R \times \mathbf{u}^R \quad \text{and} \quad \{\mathbf{u}\} = (\mathbf{u}^L + \mathbf{u}^R) / 2,
\]

respectively. Here \( \mathbf{u}^L \) and \( \mathbf{u}^R \) are the values of the trace of \( \mathbf{u} \) at \( \partial K^L \) and \( \partial K^R \), respectively. At the boundary \( \partial \Omega \), we set \( \{\mathbf{u}\} = \mathbf{u} \) and \( [\mathbf{u}]_T = \mathbf{n} \times \mathbf{u} \). We furthermore introduce the global lifting operator \( \mathcal{R}(\mathbf{u}) : [L^2(\mathcal{F}_h)]^3 \to \Sigma^p_h \) as

\[
(\mathcal{R}(\mathbf{u}), \mathbf{v})_\Omega = \int_{\mathcal{F}_h} \mathbf{u} \cdot \{\mathbf{v}\} \, dA, \quad \forall \mathbf{v} \in \Sigma^p_h, \tag{8}
\]

and, for a given face \( F \in \mathcal{F}_h \), the local lifting operator \( \mathcal{R}_F(\mathbf{u}) : [L^2(F)]^3 \to \Sigma^p_h \) as

\[
(\mathcal{R}_F(\mathbf{u}), \mathbf{v})_\Omega = \int_F \mathbf{u} \cdot \{\mathbf{v}\} \, dA, \quad \forall \mathbf{v} \in \Sigma^p_h. \tag{9}
\]

Note that \( \mathcal{R}_F(\mathbf{u}) \) vanishes outside the elements connected to the face \( F \) so that for a given element \( K \in \mathcal{T}_h \) we have the relation

\[
\mathcal{R}(\mathbf{u}) = \sum_{F \in \mathcal{F}_h} \mathcal{R}_F(\mathbf{u}), \quad \forall \mathbf{u} \in [L^2(\mathcal{F}_h)]^3. \tag{10}
\]

The discrete weak formulation for DG-FEM now reads as follows [21, 22]: Find \( \mathbf{E}_h \in \Sigma^p_h \) such that \( \forall \phi \in \Sigma^p_h \) the relation

\[
\frac{\partial^2}{\partial t^2} (\mathbf{E}_h, \phi) + \frac{\partial}{\partial t} (\sigma \mathbf{E}_h, \phi) + (\mu^{-1} \nabla_h \times \mathbf{E}_h, \nabla_h \times \phi) - \int_{\mathcal{F}_h} [\mathbf{E}_h]_T \cdot \{\nabla_h \times \phi\} \, dA - \int_{\mathcal{F}_h} \{\nabla_h \times \mathbf{E}_h\} \cdot [\phi]_T \, dA + \sum_{F \in \mathcal{F}_h} C_F (\mathcal{R}_F([\mathbf{E}_h]_T), \mathcal{R}_F([\phi]_T))_\Omega = - \left( \frac{\partial J}{\partial \phi}, \phi \right) \tag{11}
\]

is satisfied, where the operator \( \nabla_h \) denotes the elementwise application of \( \nabla \). The constant \( C_F \) is independent of both the polynomial order and the mesh size, and see [21, 22] on how to choose it.

Again, (11) is satisfied if and only if it is satisfied for every basis function \( \psi_i, i = 1, \ldots, N \), with \( N \) being the global number of degrees of freedom. Substitution of \( \mathbf{E} \approx
\[ E_h = \sum_i u_i(t) \psi_i(x) \] into (11) yields the semi-discrete system (2) with

\[
[M_e]_{ij} = (\varepsilon \psi_i, \psi_j), \quad [M_\sigma]_{ij} = (\sigma \psi_i, \psi_j), \quad [j]_i = -\left( \frac{\partial J}{\partial t}, \psi_i \right),
\]

\[
[S_\mu]_{ij} = (\mu^{-1}_r \nabla \times \psi_i, \nabla \times \psi_j) - \int_{F_h} \left[ \psi_i \right]_T \cdot \left\{ \left[ \nabla \times \psi_j \right] \right\} dA
\]

\[
- \int_{F_h} \left\{ \nabla \times \psi_i \right\} \cdot \left[ \psi_j \right]_T dA + \sum_{F \in F_h} C_F \left( R_F \left( \left[ \psi_i \right]_T \right), R_F \left( \left[ \psi_j \right]_T \right) \right)_\Omega.
\]

The matrices \( M_e \) and \( M_\sigma \) are now block-diagonal with the elementwise matrices being the blocks. However, the stiffness matrix \( S_\mu \) has still many entries far off the diagonal because of the face integrals in its construction. That is why, DG-FEM in general warrants the use of explicit time-integration schemes but not implicit ones. Even in the presence of significant stiffness, implicit methods are best avoided and local time-stepping used instead [23, 24, 25].

We emphasise that (11) is only one of many possible formulations of DG-FEM, depending on the numerical flux one chooses to use. The one we have introduced here is based on the numerical flux from [26] (see also [27]), and was analysed in detail for the time-harmonic Maxwell equations in [21, 22]. See also [20] for an overview of DG-FEM methods for elliptic problems and for a large number of possible choices for the numerical flux.

2.3. The energy norm

Convergence results for FEMs are generally derived not only in the \( L^2 \)-norm – induced by the inner product (3) – but also in a norm associated with the discrete energy of the approximation [28, 29]. These are defined for the \( H(\text{curl}) \)-conforming and DG discretisations as

\[ \| v \|_{H(\text{curl})}^2 = \| v \|^2 + \| \nabla \times v \|^2 \]

and

\[ \| v \|_{DG}^2 = \| v \|^2 + \| \nabla_h \times v \|^2 + \| h^{-\frac{1}{2}} [ v ]_T \|^2_{F_h}, \]

respectively. In the above definition, \( \| \cdot \|_{F_h} \) denotes the \( L^2(F) \) norm and \( h(x) = h_F \) is the diameter of face \( F \) containing \( x \). We note that the two definitions of the energy norm are actually identical as \( \nabla_h \) becomes \( \nabla \) and \( [ \cdot ]_T \) vanishes if the \( H(\text{curl}) \)-conforming discretisation is used.

3. Stability of the semi-discrete system

To carry out a basic stability analysis, we first transform (2) into a first-order system of ODEs,

\[ u' = v, \]

\[ M_e v' + M_\sigma v + S_\mu u = j. \]
Recall that $S_\mu$ is symmetric and therefore – using the inner-product notation for discrete vectors – we have the property

$$\frac{d}{dt}(v^T M_v + u^T S_\mu u) = \frac{dv^T}{dt} M_v v + v^T M_v \frac{dv}{dt} + \frac{du^T}{dt} S_\mu u + u^T S_\mu \frac{du}{dt} = 2v^T (-M_\sigma v - S_\mu u + j) + 2v^T S_\mu u = 2v^T j - 2v^T M_\sigma v. \quad (13)$$

If $j = 0$, this entails stability, that is

$$\frac{d}{dt}(v^T M_v + u^T S_\mu u) = -2v^T M_\sigma v \leq 0,$$

since, for constant $\sigma$, the matrix $M_\sigma$ is positive definite if $\sigma > 0$ and $M_\sigma = 0$ if $\sigma = 0$. Therefore, if $\sigma = 0$ in addition to $j = 0$, (13) shows conservation.

In order to use a stability test model introduced later in this section, we transform (12) to an equivalent explicit form. To do so, we multiply the first equation in (12) with $M_v$ and introduce the Cholesky factorisation $LL^T = M_v$. The new variables $\tilde{v} = L^T v$ and $\tilde{u} = L^T u$ then satisfy the system

$$\begin{pmatrix} \ddot{v} \\ \dot{v} \end{pmatrix} = \begin{pmatrix} 0 & I \\ \tilde{S}_\mu & \tilde{M}_\sigma \end{pmatrix} \begin{pmatrix} \dot{u} \\ \tilde{v} \end{pmatrix} + \begin{pmatrix} 0 \\ j \end{pmatrix}, \quad (14)$$

where

$$\tilde{j} = L^{-1} j, \quad \tilde{S}_\mu = L^{-1} S_\mu L^{-T}, \quad \tilde{M}_\sigma = L^{-1} M_\sigma L^{-T}.$$ 

Since both the conductivity coefficient $\sigma$ and the permittivity coefficient $\varepsilon_r$ are constant scalars in (1), the matrix $\tilde{M}_\sigma$ in (14) is the constant diagonal matrix

$$\tilde{M}_\sigma = \gamma I, \quad \gamma = \frac{\sigma}{\varepsilon_r}.$$ 

From this we can derive a two-by-two system through which stability of time-integration methods for (12) can be examined.

The matrix $\tilde{S}_\mu$ is symmetric positive semi-definite so it can be decomposed as $\tilde{S}_\mu = UAU^T$, where $A$ is a diagonal matrix with the eigenvalues of $\tilde{S}_\mu$ on its diagonal

$$\lambda_1 \geq \lambda_2 \geq \ldots \geq \lambda_r \geq \lambda_{r+1} = \lambda_{r+1} = \ldots = \lambda_n = 0,$$

where $r$ is the rank of the matrix. The matrix $U$ is orthogonal and its columns are the eigenvectors of $\tilde{S}_\mu$. Using a permutation matrix $P$, we have

$$A = \begin{pmatrix} 0 & I \\ -\tilde{S}_\mu & -\tilde{M}_\sigma \end{pmatrix} = \begin{pmatrix} 0 & U U^T \\ -U A U^T & -\gamma I \end{pmatrix} = \begin{pmatrix} U & 0 \\ 0 & U \end{pmatrix} \begin{pmatrix} 0 & I \\ -A & -\gamma I \end{pmatrix} \begin{pmatrix} U^T & 0 \\ 0 & U^T \end{pmatrix} = (U \ 0 \ 0 \ U) P \Lambda P^T \begin{pmatrix} U^T & 0 \\ 0 & U^T \end{pmatrix}, \quad (15)$$
where $A_P$ is a block-diagonal matrix with two-by-two blocks
\[
\begin{pmatrix}
0 & 1 \\
-\lambda_k & -\gamma
\end{pmatrix}, \quad k = 1, \ldots, N. \tag{16}
\]

This allows us to state the following proposition.

**Proposition 1.** Assume that $\sigma$ and $\varepsilon$, are scalar and $\gamma = \sigma/\varepsilon$. Then the matrix $A$ has

(i) $n-r$ zero eigenvalues,

(ii) $n-r$ eigenvalues that equal $-\gamma$,

(iii) $2r$ eigenvalues that are
\[
\frac{-\gamma \pm \sqrt{\gamma^2 - 4\lambda_k}}{2}, \quad k = 1, \ldots, r.
\]

Thus, the orthogonal transformation $V \equiv \begin{pmatrix} U & 0 \\ 0 & 0 \end{pmatrix} \mathcal{P}$ decouples (14) into $r$ two-by-two systems
\[
\begin{pmatrix}
\hat{u}' \\
\hat{v}'
\end{pmatrix} = \begin{pmatrix}
0 & 1 \\
-\lambda & -\gamma
\end{pmatrix} \begin{pmatrix}
\hat{u} \\
\hat{v}
\end{pmatrix} + \begin{pmatrix}
0 \\
0
\end{pmatrix},
\]

with $\lambda = \lambda_k > 0$, $k = 1, \ldots, r$, and $n-r$ two-by-two systems
\[
\begin{pmatrix}
\hat{u}' \\
\hat{v}'
\end{pmatrix} = \begin{pmatrix}
0 & 1 \\
0 & -\gamma
\end{pmatrix} \begin{pmatrix}
\hat{u} \\
\hat{v}
\end{pmatrix} + \begin{pmatrix}
0 \\
0
\end{pmatrix}.
\]

For the stability analysis, we may neglect the source term and thus arrive at the two-by-two stability test model
\[
\begin{pmatrix}
\hat{u}' \\
\hat{v}'
\end{pmatrix} = \begin{pmatrix}
0 & 1 \\
-\lambda & -\gamma
\end{pmatrix} \begin{pmatrix}
\hat{u} \\
\hat{v}
\end{pmatrix}, \quad \lambda \geq 0, \quad \gamma \geq 0. \tag{17}
\]

The attractive feature of this formulation is that stability for the test model (17) is necessary and sufficient for the stability of (12) in the norm generated by the inner product in (13).

**4. Time-integration methods**

Probably the most popular time-integration methods to use in combination with high-order DG methods are high-order Runge-Kutta methods, giving rise to what are collectively called the Runge-Kutta DG (RKDG) methods [5]. For continuous and $H(\text{curl})$-conforming FEMs geometric integrators are also widely used thanks to their ability to conserve symplecticity\(^2\) at the discrete level [11]. In this section, we briefly recall the construction of these two families of methods and we also discuss local and global Richardson extrapolation.

\(^2\)The preservation of symplecticity is important because it is related to energy. More precisely, for symplectic integrators the error in total energy will remain within a certain margin throughout the entire time integration.
The highest-order polynomial we use within the finite element methods is \( p = 3 \). For both the DG and the \( H(\text{curl}) \)-conforming methods, this corresponds to fourth-order convergence for the semi-discrete system (2) provided that the solution is smooth [28, 30, 31, 17]. Therefore, we now only discuss time-integration methods that are also at most fourth-order accurate. Extension to higher order, however, is usually straightforward.

For investigating the properties of any given time-integration method, let \( \tau \) denote the time-step size and introduce \( z_\lambda = \tau \sqrt{\lambda} \) and \( z_\gamma = \tau \gamma \).\(^3\) The stability of the time-integration method can then, in general, be best inspected through the (numerically determined) stability region

\[
S = \{(z_\lambda, z_\gamma) : z_\lambda, z_\gamma \geq 0 \text{ with } |\mu| < 1, \mu \text{ eigenvalues of the amplification operator}\}
\]

associated with the test model (17).

4.1. Runge-Kutta methods

Out of the many different types of Runge-Kutta methods, strong-stability-preserving Runge-Kutta methods (SSPRK) [3] are particularly well suited for the time integration of semi-discrete hyperbolic problems.

With the definition of the initial values \( U_0 = u_n \) and \( V_0 = v_n \) for the time step from \( t_n \) to \( t_{n+1} \), the general \( s \)-stage SSPRK scheme for (12) reads

\[
U_k = \sum_{l=0}^{k-1} (\alpha_{kl} U_l + \tau \beta_{kl} V_l),
\]

\[
M_s V_k = \sum_{l=0}^{k-1} (\alpha_{kl} V_l + \tau \beta_{kl} (-S_p U_l - M_s V_l + j(t_l))),
\]

\[
U_{n+1} = U_s, \quad V_{n+1} = V_s,
\]

where \( k = 1, \ldots, s \), while \( \alpha_{kl} \) and \( \beta_{kl} \) are the coefficients in the SSPRK method. Applying (18) to the test equation (17), the amplification operator \( \mathcal{M}_{\text{ssp}}^k \) of an \( s \)-stage SSPRK method is

\[
\mathcal{M}_{\text{ssp}}^k = \sum_{l=1}^{k-1} B_{kl} \mathcal{M}_{\text{ssp}}^{l-1} \quad \text{with} \quad B_{kl} = \alpha_{kl} \begin{pmatrix} 1 & 0 \\ 0 & 1 \end{pmatrix} + \beta_{kl} \begin{pmatrix} 0 & 1 \\ z_\lambda^{-2} & -z_\gamma \end{pmatrix},
\]

where, again, \( k = 1, \ldots, s \), and \( \mathcal{M}_{\text{ssp}}^0 \) is the identity. We show the stability regions of several SSPRK schemes in Figure 2, where we refer to an \( s \)-stage \( p \)-th-order SSPRK method as SSPRK\((s,p)\). It is important to emphasise that any (standard as well as ‘non-standard’ such as SSP) explicit \( s \)-stage \( p \)-th-order RK methods with \( s = p \) has the same amplification matrix. In those cases the choice of coefficients only determines the SSP property and not the shape of the stability region. For the cases when \( s = p + 1 \), we

\(^3\) These values appear in a natural way in the amplification matrices of most time-integration methods described later in this section.
display the stability regions of the methods that were derived and analysed in [32, 33, 34]. The plots suggest that increasing the number of stages, while keeping the polynomial order fixed, results in a more favourable time-step restriction for the conduction part – one that more than offsets the cost of introducing an additional stage at each time step. This is in line with known results for the linear advection equation [32, 33]. Nevertheless, explicit SSPRK methods treat the conduction term, as well as the wave term, explicitly, which entails a time-step condition that is too restrictive even for moderately conductive materials (see next section). Note that the second-order methods are only stable for \( z_\gamma > 0 \), i.e. for \( \sigma > 0 \).

4.2. Composition methods

Composition methods [35, 36, 37] are especially suitable for geometric integration [11] and thus for the time integration of first-order Hamiltonian systems. Our description of the composition methods here strictly follows that in [10] and we refer to that work for more details.

The second-order composition method (CO2) for (12) is defined as

\[
\frac{u_{n+1/2} - u_n}{\tau} = \frac{1}{2} v_n, \\
M_\sigma \frac{v_{n+1} - v_n}{\tau} = -S_\mu u_{n+1/2} - \frac{1}{2} M_\sigma (v_n + v_{n+1}) + \frac{1}{2} \left( j(t_n) + j(t_{n+1}) \right),
\]

which is akin to the ubiquitous leapfrog scheme, with the only difference being in the treatment of the source term (cf. [38]). If applied to the test model (17), it has the amplification matrix

\[
\mathcal{M}_{\text{CO2}} = \begin{pmatrix}
1 - \frac{1}{2} z_\lambda^2 & \frac{1}{2} z_\gamma \\
-\frac{1}{2} z_\lambda^2 & 1 - \frac{1}{2} z_\lambda^2 - \frac{1}{2} z_\gamma
\end{pmatrix},
\]

which entails the stability properties: \( z_\lambda \leq 2 \) if \( z_\gamma = 0 \) and \( z_\lambda < 2 \) if \( z_\gamma > 0 \). An attractive feature of this method over explicit RK methods is that it is unconditionally stable with respect to the conduction term.

In principle, it is possible to construct an arbitrary high-order composition method [35]. In this article, however, we are only interested in at most fourth-order accurate methods so we will now only discuss the fourth-order composition method (CO4). We define the initial values for the inner time step as \( U_0 = u_n \) and \( V_0 = v_n \), time levels \( t^u, t^v \) for \( u, v \) and coefficients

\[
\beta_0 = \alpha_0 = 0, \quad \beta_1 = \alpha_5 = \frac{14 - \sqrt{19}}{108}, \quad \beta_2 = \alpha_4 = \frac{-23 - 20\sqrt{19}}{270}, \\
\beta_3 = \alpha_3 = \frac{1}{3}, \quad \beta_4 = \alpha_2 = \frac{-2 + 10\sqrt{19}}{135}, \quad \beta_5 = \alpha_1 = \frac{146 + 5\sqrt{19}}{540}.
\]

11
Figure 2: Stability regions (shaded areas) for several explicit SSPRK (s,p) methods, where s is the number of stages and p is the order of the method. Note that all explicit RK methods with s = p have the same stability regions as SSPRK(s,p) with s = p (left column).
The fourth-order composition method \cite{35, 36, 10} for (12) now reads
\[
\frac{U_k - U_{k-1}}{\tau} = (\beta_k + \alpha_{k-1}) V_{k-1},
\]
\[
M_s \frac{V_k - V_{k-1}}{\tau} = \beta_k \left( -S_\mu U_k - M_\sigma V_{k-1} + f(t_k^s) \right) + \alpha_k \left( -S_\mu U_k - M_\sigma V_k + f(t_k^s) \right),
\]
where \( k = 1, \ldots, s \), \( s = 5 \) is the number of internal time levels, and \( t_k^s = t_n + (\tilde{\alpha}_k + \tilde{\beta}_k) \tau \) and \( t_k^* = t_n + (\tilde{\alpha}_{k-1} + \tilde{\beta}_k) \tau \) with the coefficients \( \tilde{\alpha}_k = \alpha_1 + \cdots + \alpha_k \) and \( \tilde{\beta}_k = \beta_1 + \cdots + \beta_k \).

The amplification operator of (22) when applied to (17) is then
\[
\prod_{k=5}^{1} \frac{1}{1 + \alpha_k z_{\gamma}} \left( \frac{1 + \alpha_k z_{\gamma}}{1 - \beta_k z_{\gamma} - (\beta_k + \alpha_k) z_{\lambda}^2} \right), \quad (23)
\]
An important property of any fourth-order composition method is that it inevitably contains a negative coefficient, which in our case is \( \alpha_4 = \beta_2 \). This entails a stability restriction that is conditional even for an implicitly treated conduction term. This is illustrated in Figure 3, where parts of the upper right half of the stability region for (22) is shown. Stability is guaranteed as long as \( z_{\gamma} < 2.4 \) and \( z_{\lambda} < 3 \), or equivalently, if \( \tau < 2.4/\gamma \) and \( \tau < 3/\sqrt{\lambda} \).

4.3. Fourth-order Richardson extrapolation

As already mentioned in the previous section, when \( \sigma > 0 \) the stability condition may be very restrictive even for moderately conductive materials. In these cases, high-order composition methods and SSPRK methods are not competitive. Instead, one would prefer to use explicit methods which treat the conduction term in an unconditionally stable manner. Since the second-order composition method (20) is such a method, extending it to higher order through Richardson extrapolation is an obvious alternative. We refer
Figure 4: Stability region (shaded area) of the fourth-order local Richardson extrapolation (25). The right plot zooms in on the region where stability for wave term \( z_\lambda \) is guaranteed. Stability for the conduction term \( z_\gamma \) is unconditional.

to [10] for a detailed discussion on the stability properties of the fourth-order local and global versions of the Richardson extrapolation. Here we first recall the construction of the fourth-order global Richardson extrapolation (GEX4)

\[
  u_{T}^{\text{gex4}} = \frac{4}{3} u_{T/2}^{\text{co2}} - \frac{1}{3} u_{T}^{\text{co2}},
\]

(24)

where \( u_{T}^{\text{co2}} \) and \( u_{T/2}^{\text{co2}} \) denote the results at final time computed by the second-order composition method with time steps \( \tau \) and \( \frac{\tau}{2} \), respectively. Since extrapolation only takes place once at the final time of the integration, this method has the same stability properties as the second-order composition method. Note that it only needs three times as much computational work per time step.

For long time integration and in the absence of damping, global extrapolation may not be sufficiently effective in annihilating leading error terms. In these cases, the local version of Richardson extrapolation – when the extrapolation is performed at each time step – is usually more beneficial. The local version of (24) is, however, not unconditionally stable with respect to \( z_\gamma \). Instead, we can use the fourth-order local extrapolation (LEX4) defined as

\[
  u_{T}^{\text{lex4}} = \frac{9}{8} u_{T/3}^{\text{co2}} - \frac{1}{8} u_{T}^{\text{co2}},
\]

(25)

where the work per time step is approximately four times as much as that of CO2. The amplification operator of LEX4 for the test model (17) reads

\[
  \frac{9}{8} \mathcal{M}_{\text{co2}}^{3}(z_\lambda/3, z_\gamma/3) - \frac{1}{8} \mathcal{M}_{\text{co2}}(z_\lambda, z_\gamma),
\]

(26)

where \( \mathcal{M}_{\text{co2}}(z_\lambda, z_\gamma) \) denotes the amplification operator (21) of CO2. Figure 4 shows the associated stability region \( S \), which indicates an approximate stability interval \( 0 \leq z_\lambda \leq 2.85 \) and unconditional stability for \( z_\gamma \).
5. Convergence and comparison of performance

In this section, we use a simple test example to illustrate the numerical performance of the two spatial discretisation techniques described in Section 2. For both methods, the predicted convergence rate of the semi-discrete system is $O(h^{p+1})$ in the $L^2(\Omega)$ norm and $O(h^p)$ in the energy norm for smooth solutions [17, 39, 21, 28, 30]. It is thus natural to choose the time-integration method such that it guarantees at least the same order of convergence. Therefore, if the polynomial order in the FEM is at most one we use the second-order composition method (20); if the polynomial order is two or three we apply one of the possible fourth-order methods described in Section 4.

The numerical tests are implemented in $hp$GEM [40], a general finite element package suitable for solving a variety of physical problems in fluid dynamics and electromagnetism. To integrate the semi-discrete system in time we use PETSc [41], where we set the tolerance at $\text{tol} = 10^{-8}$ as a stopping criterion in the linear solver for the $H(\text{curl})$-conforming method.

In the example, we consider (1) in the cubic domain $\Omega = (0, 1)^3$. We define the time-independent field

$$\bar{E}(x, y, z) = \begin{pmatrix} \sin(\pi y) \sin(\pi z) \\ \sin(\pi y) \sin(\pi x) \\ \sin(\pi x) \sin(\pi y) \end{pmatrix}$$

and choose the source term to be

$$\frac{\partial J}{\partial t} = (\varepsilon \eta''(t) + \sigma \eta'(t) + 2\pi^2 \eta(t)) \bar{E}(x, y, z).$$

The exact solution then reads

$$\mathbf{E}(t, x, y, z) = \eta(t) \bar{E}(x, y, z), \quad \eta(t) = \sum_{k=1}^{3} \cos k \omega_k t,$$

with $\omega_1 = 1$, $\omega_2 = 1/2$, $\omega_3 = 1/3$, $\varepsilon_1 = 1$.

When the globally $H(\text{curl})$-conforming discretisation [17, 31] is used, we need to solve a linear system at each time step. Since the matrix $M_t$ is positive definite, a natural choice of linear solver is the preconditioned conjugate gradient (PCG) method. We apply the incomplete Cholesky preconditioner for all meshes and polynomial orders. While more advanced preconditioning techniques certainly do exist, notably multigrid [42, 43, 44, 45], it is well beyond the scope of this paper to investigate them for high-order tetrahedral finite-element meshes. In any case, results with PCG preconditioner are intended as benchmark calculations only.

5.1. Convergence of the damped system

As a first example, we run (27) until final time $T_{\text{end}} = 12\pi$ on a sequence of structured meshes with $N_{\text{el}} = 5, 40, 320, 2560, 20480, 163840$ elements. The conductivity is prescribed as $\sigma = 60\pi$, which corresponds to the dimensional value $\bar{\sigma} = 0.58 \text{ m}^{-1}$, typical of the

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4 The software is free to download and available at [www.math.utwente.nl/~hpgemdev/download.html](http://www.math.utwente.nl/~hpgemdev/download.html).
human abdomen. In each mesh the largest face diameter \( h \) is exactly half that of the previous mesh. We plot the convergence rates in Figure 5 in both the \( L^2(\Omega) \)-norm and the energy norm for polynomial orders \( p = 1, 2, 3 \).

We can see that the expected convergence rates are achieved asymptotically for both the DG and the \( H(\text{curl}) \)-conforming methods, and that it takes fewer degrees of freedom for the \( H(\text{curl}) \)-conforming discretisation to reach a given accuracy. We can also confirm the well-established observation that the use of high-order approximations pays off (at least for smooth solutions) in terms of accuracy per degrees of freedom. The results for the conservative system \( \sigma = 0 \) are very similar and therefore omitted from this article.

5.2. Comparison of performance

To gain further insight into the computational costs of the time integration, we show the performance of the DG method in Table I and that of the \( H(\text{curl}) \)-conforming method in Table II. In this particular example, we use a structured mesh with 320 elements and an unstructured one with 432 elements.\(^5\) Although the accuracy of the two methods is comparable, the computational costs are not and the pattern changes dramatically as the order increases. The total number of matrix-vector multiplications (matvecs) needed to integrate until \( T_{\text{end}} \) is always higher for the \( H(\text{curl}) \)-conforming case than for the DG method. This is not surprising given that at each time step a linear system has to be solved. However, this seemingly unfavourable property does not manifests itself in longer computational time for \( p = 1 \) and \( p = 2 \) on structured meshes, thanks in part to the smaller size of the system and in part to a weaker time-step restriction in the \( H(\text{curl}) \)-conforming FEM. The situation is different for \( p = 3 \). Here, the increased number of matvecs translates readily into more CPU time. The effect is even more pronounced on unstructured meshes, where DG performs slightly better for \( p = 2 \) already and where the \( H(\text{curl}) \)-conforming computation for \( p = 3 \) is excessively long.

5.3. The effect of conductivity on the performance

The role of the conductivity can be best assessed by comparing the performance of the different high-order time-integration methods for \( p = 3 \). In the cases of CO4, the conduction term poses a stricter time-step size than the wave term and increases the number of time steps and thus the computational cost. On the structured mesh with 320 elements and \( \sigma = 60\pi \), this only affects the \( H(\text{curl}) \)-conforming discretisation because the stiffness matrix in the DG method has a significantly larger spectral radius (and therefore it still determines the stability condition). On the unstructured mesh with 432 elements and \( \sigma = 450\pi \), however, it affects the DG discretisation, too. We note that the results for the SSPRK method are similar to those for CO4 and to what one would expect from the stability regions in Figure 2 and therefore not reported.

This result indicates that physically feasible values of \( \sigma \) may prohibit the use of fourth-order (or, indeed, any high-order) composition methods, as well as explicit RK methods. Instead, Richardson extrapolation based on the second-order composition method may be used since they are unconditionally stable with respect to the conductivity term.

\(^5\)A mesh of 320 or 432 tetrahedra is sufficient to compare the different methods from the point of view of accuracy and computational work. A finer mesh would naturally give a more accurate solution but the relative performance of the methods would remain the same.
Figure 5: Convergence plots in the $L^2$-norm (left column) and in the energy norm (right column) for test example (27) with $\sigma = 60\pi$. In each plot the convergence rates of the DG method and the $H(\text{curl})$-conforming method are shown along with the expected order of convergence.
Table I: Computational costs of the DG method for example (27) with $\sigma = 60\pi$. A structured mesh of 320 elements (top table) and an unstructured mesh of 432 elements (bottom table) are used with spatial polynomial orders $p = 1, 2, 3$.

<table>
<thead>
<tr>
<th>Structured</th>
<th>method</th>
<th># DoF</th>
<th>$L^2(\Omega)$ error</th>
<th># matvecs</th>
<th>$\tau$</th>
<th>CPU time</th>
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</tr>
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Table II: Computational costs of the $H(\text{curl})$-conforming method for example (27) with $\sigma = 60\pi$. A structured mesh of 320 elements (top table) and an unstructured mesh of 432 elements (bottom table) are used with spatial polynomial orders $p = 1, 2, 3$.

<table>
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<td>0.01530</td>
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6. Numerical dispersion analysis of DG-FEM

In this section, we carry out a numerical dispersion analysis of the semi- and fully discrete system with DG spatial discretisation. This is done in the following way: i) solve the time-harmonic eigenvalue problem, which corresponds to the semi-discrete system
with Fourier mode initial conditions; ii) apply a chosen time-integration method to the test model (17) with the computed semi-discrete numerical frequency. This approach has two main advantages over simply solving the eigenvalue problem that results from applying the amplification matrix directly to (12). First, it is more efficient because we solve an eigenvalue problem that is smaller and always symmetric. Second, it makes it possible to study the dispersion (and dissipation) properties of the time-integration scheme separately from those of the semi-discrete scheme. To investigate the dispersion and dissipation properties of the fully discrete schemes, we consider the semi-discrete system (12) with \( \sigma = 0 \) and \( j = 0 \),

\[
\begin{pmatrix}
    u' \\
    v'
\end{pmatrix} = \mathcal{A} \begin{pmatrix}
    u \\
    v
\end{pmatrix} \quad \text{with} \quad \mathcal{A} = \begin{pmatrix}
    0 & I \\
    -M^{-1}c & 0
\end{pmatrix},
\]

and assume a plane wave exact solution

\[
\mathbf{E}(x, t) = \hat{\mathbf{E}} \exp(-i\omega t) \exp(i\mathbf{k} \cdot \mathbf{x})
\]

with periodic boundary conditions and \( \hat{\mathbf{E}} = 1 \). In (29), \( i^2 = -1 \), \( \omega \) denotes the angular frequency, and \( k = (k_x, k_y, k_z)^T \) is the wave number. Between these quantities the (exact) dispersion relation \( \omega^2 = k^2/c^2 \) holds with \( k^2 = k_x^2 + k_y^2 + k_z^2 \) and with \( c = 1/\sqrt{\varepsilon_r \mu_r} \), which is the speed of light.

As a first step, we project the exact initial conditions \( \mathbf{E}(x, 0) \) and \( \partial_t \mathbf{E}(x, 0) \) onto the finite-element space

\[
\begin{align*}
E_h^j(0) &= (\mathbf{E}(x, 0), \psi_j)_\Omega, & j &= 1 \ldots N, \\
\frac{d}{dt}E_h^j(0) &= (\partial_t \mathbf{E}(x, 0), \psi_j)_\Omega, & j &= 1 \ldots N.
\end{align*}
\]

We can now obtain the initial conditions for (28) through the relations \( u_0 = u(0) = M^{-1}\varepsilon E_h(0) \) and \( v_0 = v(0) = u'(0) = M^{-1}c \partial_t E_h(0) \). The time-exact discrete Fourier mode at time level \( n\tau \) is then defined as

\[
\begin{pmatrix}
    u_n \\
    v_n
\end{pmatrix} = \nu^n \begin{pmatrix}
    u_0 \\
    v_0
\end{pmatrix} \quad \text{with} \quad \nu^n = e^{-i\omega_h n\tau},
\]

where \( \nu^n \) is the exact amplification factor and \( \omega_h \) is the semi-discrete numerical frequency.

6.1. Dispersion analysis of the semi-discrete system

To investigate the impact of the space discretisation only, we consider the semi-discrete equation

\[
M_c u'' + S_\mu u = 0
\]

with periodic boundary conditions and a plane wave initial condition (29). In this case, (32) is equivalent to the discrete time-harmonic Maxwell eigenvalue problem

\[
S_\mu u - \omega_h^2 M_c u = 0
\]
Table III: Semi-discrete frequencies $\omega_h$ (top table) and corresponding frequency errors $\omega_h - \omega_{\text{ex}}$ (bottom table) of the DG method for exact frequency $\omega_{\text{ex}} = \sqrt{8}\pi$.

<table>
<thead>
<tr>
<th>$p$</th>
<th>$h = \frac{1}{2}$</th>
<th>$h = \frac{1}{4}$</th>
<th>$h = \frac{1}{8}$</th>
<th>$h = \frac{1}{16}$</th>
<th>$\omega_{\text{ex}}$</th>
</tr>
</thead>
<tbody>
<tr>
<td>1</td>
<td>9.4286</td>
<td>9.0469</td>
<td>8.9271</td>
<td>8.8858</td>
<td></td>
</tr>
<tr>
<td>2</td>
<td>9.4738</td>
<td>8.9276</td>
<td>8.8887</td>
<td></td>
<td>8.8858</td>
</tr>
<tr>
<td>3</td>
<td>8.9146</td>
<td>8.8875</td>
<td></td>
<td></td>
<td>8.8858</td>
</tr>
</tbody>
</table>

<table>
<thead>
<tr>
<th>$h = \frac{1}{2}$</th>
<th>$h = \frac{1}{4}$</th>
<th>$h = \frac{1}{8}$</th>
<th>$h = \frac{1}{16}$</th>
<th>$\omega_{\text{ex}}$</th>
</tr>
</thead>
<tbody>
<tr>
<td>$p = 1$</td>
<td>5.4283e-01</td>
<td>1.6117e-01</td>
<td>4.1380e-02</td>
<td></td>
</tr>
<tr>
<td>$p = 2$</td>
<td>5.8800e-01</td>
<td>4.1831e-02</td>
<td>2.9628e-03</td>
<td></td>
</tr>
<tr>
<td>$p = 3$</td>
<td>2.8869e-02</td>
<td>1.7173e-03</td>
<td>3.0850e-05</td>
<td></td>
</tr>
</tbody>
</table>

with periodic boundary conditions. All semi-discrete eigenvalues $\omega_h^2$ of (33) are real and non-negative, which entails that the space discretisation imposes no dissipation. In Table III, we show the numerical frequencies of the spatial DG discretisation for the Fourier mode with $k_x = 2\pi, k_y = -2\pi, k_z = 0$, i.e. with exact angular frequency $\omega_{\text{ex}} = \sqrt{8}\pi$. The number of elements for each mesh is $N_0 = 5\left(\frac{1}{h}\right)^3$ and in each element the local number of degrees of freedom is $\frac{1}{2}(p+1)(p+2)(p+3)$. To solve the eigenvalue discrete problem (33) of this size the Matlab implementation of the Jacobi-Davidson iterative method [46, 47] is used. We note that for other Fourier modes the same approximation properties apply as long as $\omega_h h$ is in the same region as shown in the tables. The frequency errors for the same meshes and polynomial orders are depicted in the bottom half of Table III. Note that the frequency errors are signed, indicating phase advance.

6.2. Dispersion analysis of the fully discrete system

To include the time integration in the dispersion analysis it suffices to apply a chosen time-integration method to the test model (17) with $\gamma = 0$. We are allowed to do that because the eigenvalues of $\tilde{S}_\mu$ are the same as the eigenvalues of $M^{-1}S_\mu$, that is $\lambda = \omega_h^2$. Let $\mathcal{M}$ denote the amplification operator of any of the time-integration methods described in Section 4. So instead of (31) we now have the fully discrete Fourier mode at time level $n\tau$,

$$v_{h}^{n+1} \begin{pmatrix} u_0 \\ v_0 \end{pmatrix} = \mathcal{M} v_{h}^{n} \begin{pmatrix} u_0 \\ v_0 \end{pmatrix},$$

which reduces to the eigenvalue problem

$$v_{h} \begin{pmatrix} u_0 \\ v_0 \end{pmatrix} = \mathcal{M} \begin{pmatrix} u_0 \\ v_0 \end{pmatrix}.$$

Solving this eigenvalue problem will produce two eigenpairs, representing two waves with the same wave number but travelling in opposite directions. Without loss of generality, we can discard the one with negative real part and establish the dispersive and dissipative

---

6The software is free to download and available at http://www.math.uu.nl/people/sleijpen.
Table IV: Frequency errors imposed only by the time integration, \( \text{Re}(\omega_h^k) - \omega_h \), of the SSPRK(4, 3), CO2 and LEX4 methods for semi-discrete numerical frequencies \( \omega_h \) taken from Table III.

<table>
<thead>
<tr>
<th>Method</th>
<th>( h = \frac{1}{2} )</th>
<th>( h = \frac{1}{4} )</th>
<th>( h = \frac{1}{8} )</th>
<th>( h = \frac{1}{16} )</th>
</tr>
</thead>
<tbody>
<tr>
<td>SSPRK(4, 3)</td>
<td>p = 1</td>
<td>—</td>
<td>7.1799e-05</td>
<td>3.6525e-06</td>
</tr>
<tr>
<td></td>
<td>p = 2</td>
<td>1.5242e-04</td>
<td>7.0867e-06</td>
<td>4.3347e-07</td>
</tr>
<tr>
<td></td>
<td>p = 3</td>
<td>2.9293e-05</td>
<td>1.8039e-06</td>
<td>1.1265e-07</td>
</tr>
<tr>
<td>CO2</td>
<td>h = \frac{1}{2}</td>
<td>—</td>
<td>9.7283e-03</td>
<td>2.1439e-03</td>
</tr>
<tr>
<td></td>
<td>p = 1</td>
<td>—</td>
<td>9.7283e-03</td>
<td>2.1439e-03</td>
</tr>
<tr>
<td></td>
<td>p = 2</td>
<td>1.4229e-02</td>
<td>2.9674e-03</td>
<td>7.3172e-04</td>
</tr>
<tr>
<td></td>
<td>p = 3</td>
<td>6.0353e-03</td>
<td>1.4930e-03</td>
<td>3.7292e-04</td>
</tr>
<tr>
<td>LEX4</td>
<td>h = \frac{1}{2}</td>
<td>—</td>
<td>-7.9554e-06</td>
<td>-4.0558e-07</td>
</tr>
<tr>
<td></td>
<td>p = 1</td>
<td>—</td>
<td>-7.9554e-06</td>
<td>-4.0558e-07</td>
</tr>
<tr>
<td></td>
<td>p = 2</td>
<td>-1.6866e-05</td>
<td>-7.8671e-07</td>
<td>-4.8152e-08</td>
</tr>
<tr>
<td></td>
<td>p = 3</td>
<td>-3.2488e-06</td>
<td>-2.0035e-07</td>
<td>-1.2516e-08</td>
</tr>
</tbody>
</table>

The properties of the fully discrete scheme through the relation

\( \nu_h = e^{-i \omega_h^k \tau}, \)

where \( \omega_h^k \) represents the fully discrete numerical frequency. The real part of \( \omega_h^k \) defines the actual angular frequency in the discrete dispersion relation, while a negative imaginary part indicates numerical dissipation. A non-negligible positive imaginary part would mean instability.

We show the frequency errors of the time-integration schemes SSPRK(4, 3), CO2 and LEX4 in Table IV. They show that the frequency error of the time-integration method is at least an order smaller than the one of the DG method, as long as the order of the time-integration method is on a par with the order of the DG method. When this is not the case, such as when CO2 is used for \( p = 2 \) or \( p = 3 \), the frequency error of the time integration is commensurate with, or exceeds that of the DG discretisation.

Composition methods, such as CO2 and CO4, are known to be non-dissipative [35]. Thus combining them with a symmetric spatial discretisation results in an energy-conservative fully-discrete discretisation. Global Richardson extrapolation based on a composition method naturally inherits this property. However, local Richardson extrapolation may introduce a slight dissipation even when based on a non-dissipative scheme such as CO2. We show this in the top half of Table V and note that the error is generally too small to have a real impact on simulations arising in practice. By comparison, the SSPRK(4, 3) scheme introduces a much more significant level of dissipation, shown in the bottom half of Table V.

Finally, we note that if a time-dependent boundary condition is used in (1) instead of a homogeneous one, order reduction may occur. See [10] for the possible effects of this.
Table V: Imaginary part of the numerical frequency, $\text{Im}(\omega^h_\tau)$, for the LEX4 and SSPRK(4, 3) time-integration method, where the semi-discrete numerical frequencies $\omega^h$ are taken from Table III. This term is responsible for numerical dissipation.

<table>
<thead>
<tr>
<th>Method</th>
<th>$h = \frac{1}{2}$</th>
<th>$h = \frac{1}{4}$</th>
<th>$h = \frac{1}{8}$</th>
<th>$h = \frac{1}{16}$</th>
</tr>
</thead>
<tbody>
<tr>
<td>LEX4</td>
<td></td>
<td>-6.9642e-07</td>
<td>-1.6998e-08</td>
<td>-4.9043e-10</td>
</tr>
<tr>
<td>p = 1</td>
<td></td>
<td>-1.7825e-06</td>
<td>-3.9053e-08</td>
<td>-1.1891e-09</td>
</tr>
<tr>
<td>p = 2</td>
<td></td>
<td>-2.3027e-07</td>
<td>-7.0689e-09</td>
<td>-2.2071e-10</td>
</tr>
<tr>
<td>p = 3</td>
<td></td>
<td>-3.8256e-04</td>
<td>-4.7337e-05</td>
<td>-5.9156e-06</td>
</tr>
<tr>
<td>SSPRK(4, 3)</td>
<td></td>
<td>-7.5911e-04</td>
<td>-8.0688e-05</td>
<td>-9.5692e-06</td>
</tr>
<tr>
<td>p = 1</td>
<td></td>
<td>-1.3346e-03</td>
<td>-1.3217e-04</td>
<td>-1.6251e-05</td>
</tr>
<tr>
<td>p = 2</td>
<td></td>
<td>-3.8256e-04</td>
<td>-4.7337e-05</td>
<td>-5.9156e-06</td>
</tr>
</tbody>
</table>

7. Concluding remarks

We have investigated the time-dependent second-order Maxwell equation in three spatial dimensions. A direct comparison between the high-order DG-FEM and the high-order $H$-(curl)-conforming FEM on both structured and unstructured meshes was provided when $H$-(curl)-conforming hierarchic basis functions are used. The computational tests have highlighted the fact that the inclusion of moderate conductivity renders many of the popular time-integration methods uncompetitive owing to a stringent time-step restriction. In these cases, global or local Richardson extrapolations based on the second-order composition method provide a viable alternative as they treat the conductivity term implicitly.

Through a numerical dispersion and dissipation analysis, we have also shown that the spatial discretisation dominates the frequency error as long as the order of the time integration is at least the same as the order of the spatial discretisation. Since the semi-discrete system is symmetric and therefore conserves (the discrete) energy, applying a composition method to integrate in time results in a fully-discrete scheme that also conserves (the discrete) energy.

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