

SAMPLED-DATA L^∞ SMOOTHING: FIXED-SIZE ARE SOLUTION WITH FREE HOLD FUNCTION*

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Abstract. The problem of estimating an analog signal from its noisy sampled measurements is studied in the L^∞ (induced L^2 -norm) framework. The main emphasis is placed on relaxing causality requirements. Namely, it is assumed that l future measurements are available to the estimator, which corresponds to the fixed-lag smoothing formulation. A closed-form solution to the problem is derived. The solution has the complexity of $\mathcal{O}(l)$ and is based on two discrete algebraic Riccati equations, whose size does not depend on the smoothing lag l .

Key words. Sampled-data systems, fixed-lag smoothing, L^∞ optimization, generalized hold functions, signal reconstruction.

1. Introduction. This paper studies the problem of estimating an analog signal v from sampled measurements of a related signal y . We assume that v and y are generated by an analog LTI system \mathcal{G} , driven by a common exogenous signal w_v as shown in Fig. 1.1. The

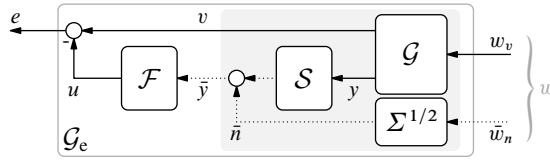


FIG. 1.1. Sampled-data estimation setup

measured discrete signal \bar{y} is the sampled version of y (\mathcal{S} denotes the ideal sampler) with a constant sampling period $h > 0$, corrupted by a discrete measurement noise \bar{n} . The latter may reflect roundoff errors and its intensity is modeled by the matrix $\Sigma = \Sigma' \geq 0$. The D/A converter \mathcal{F} (estimator), which generates an estimate u of v , is our design parameter. We quantify the estimation performance in terms of the L^∞ norm of the *error system*

$$\mathcal{G}_e := \begin{bmatrix} \mathcal{G}_v & 0 \end{bmatrix} - \mathcal{F} \begin{bmatrix} \mathcal{S}\mathcal{G}_y & \Sigma^{1/2} \end{bmatrix}, \quad (1.1)$$

which maps the aggregate exogenous signal $w := \begin{bmatrix} w_v \\ w_n \end{bmatrix}$ (see Fig. 1.1) to the estimation error $e := v - u$ (here \mathcal{G}_v and \mathcal{G}_y are the rows of \mathcal{G} corresponding to v and y , respectively). This L^∞ norm is the induced operator norm $L^2(\mathbb{R}) \times \ell^2(\mathbb{Z}) \rightarrow L^2(\mathbb{R})$.

The main theme of this study is the relaxation of causality constraints imposed upon \mathcal{F} . We say that \mathcal{F} is *l-causal* if its output $u(t)$, at a time instance $t \in \mathbb{R}$, depends only on $\bar{y}[k]$ for all $k \leq t/h + l$. In other words, an *l-causal* estimator has access to l “future” measurements of \bar{y} (l steps preview). Estimation problems in which the estimator is constrained to be *l-causal* for some $l \in \mathbb{N}$ are referred to as *fixed-lag smoothing* and l is called the *smoothing lag*, see [1, 17] and the references therein. The smoothing problem may also be interpreted as the estimation of the lh -delayed version of v by a causal estimator, so the problem is frequently referred to as the H^∞ fixed-lag smoothing, which reflects the causality of \mathcal{G}_e in this formulation.

The incentive for relaxing causality constraints on \mathcal{F} is the potential for an improved estimation performance [1]. This comes at the price of a more complex \mathcal{F} and, especially,

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a more knotty analysis compared with corresponding filtering ($l = 0$) and fixed-interval smoothing ($l = \infty$) results. Even in pure continuous- and discrete-time settings unrestricted solutions to the L^∞ (H^∞) fixed-lag smoothing problems were derived only in '00s [13, 21], more than a decade after the corresponding filtering and fixed-interval smoothing results [18, 19]. Sampled-data counterparts of these results are yet more challenging. To the best of our knowledge, there is no L^∞ fixed-lag smoothing solution for the setup in Fig. 1.1 in the literature. The filtering problem in this setting was solved in [20] in the case of $\Sigma = I$ and then in [15] for a general, possibly singular, Σ . The design of non-causal D/A converters (fixed-interval smoothing) is addressed in [9]. In the special case of $l = 1$ (and $\Sigma = 0$), [14] derives the solvability conditions, but not formulae for \mathcal{F} .

We address the sampled-data L^∞ fixed-lag smoothing problem via the lifting technique [4], which converts it to an equivalent pure discrete problem, some parameters of which are operators over infinite-dimensional spaces. We then start with a formal solution in terms of these operators and then rewrite such a solution in terms of the original parameters of \mathcal{G} . The latter procedure, called peeling-off, is rather nontrivial and its successful completion is the main technical contribution of this paper. Technical challenges of the peeling-off step in the smoothing case go far beyond those in the filtering case [15], owing chiefly to a more elaborate solution to the discrete smoothing problems.

It is well known [2, Sec. 7.3] that discrete fixed-lag smoothing can in principle be cast as a filtering ($l = 0$) problem by incorporating the delay z^{-l} in the “ v ” channel into the signal generator. This approach, however, increases the problem dimension and might blur properties of the resulting solution. In the H^2 (Kalman smoothing) case, the structure of the filtering formulae can be exploited to derive a solution that is based on fixed-size (independent of l) Riccati equation and whose computational burden is $\mathcal{O}(l)$, see [2, Sec. 7.3]. A similar approach, however, does not work so smoothly in the H^∞ optimization because the corresponding Riccati equation in this case is more involved, see [3, 7, 23] for solutions derived via this method and [21, §III-B] for a discussion about their limitations. Moreover, the application of this approach to the sampled-data problem is complicated by the fact that the “ v ” channel is intrinsically infinite dimensional in the lifted domain. To the best of our knowledge, the only complete solution to the discrete H^∞ fixed-lag smoothing problem available in the literature is the result of [21]. It provides necessary and sufficient solvability conditions and does not introduce restrictive assumptions about the signal generator. This solution, however, involves several intermediate steps, which impedes its use as a starting point for the peeling-off procedure. This motivates us to derive alternative discrete state-space formulae in [12] following the steps of [16].

The solution of [12] also involves several intermediate calculations. These calculations, however, appear to be more suitable for the use in sampled-data applications. As a result, in the current paper we succeed in deriving a numerically tractable and transparent solution to the L^∞ sampled-data problem. Our solution is based on two discrete algebraic Riccati equations, which are independent of the smoothing lag l and one of which does not depend on the achievable performance level γ either. Similarly to other sampled-data H^∞ solutions [4], our solvability conditions involve the verification of the non-singularity of a matrix function built upon blocks of a matrix exponential over the whole interval $(0, h]$. This part is the most involved numerically part of the solvability conditions. The others are just plain conditions based on the corresponding H^∞ Riccati equation. The suboptimal solution is then the cascade of a discrete filter and a zero-order generalized hold. The latter actually coincides with the D/A part of the optimal L^2 solution of [10].

Notation. For any set \mathbb{A} , its indicator function $\mathbb{1}_{\mathbb{A}}(t)$ is 1 if $t \in \mathbb{A}$ and is zero elsewhere. The space $L^2(\mathbb{R})$ is the set of functions $f : \mathbb{R} \rightarrow \mathbb{C}^{n_f}$ that have finite norm $\|f\|_2 :=$

$(\int_{t \in \mathbb{R}} \|f(t)\|^2 dt)^{1/2}$, where $\|\cdot\|$ denotes the standard Euclidean norm. $\ell^2(\mathbb{Z})$ is the set of $\tilde{f} : \mathbb{Z} \rightarrow \mathbb{C}^{n_f}$ with finite norm $\|\tilde{f}\|_2 := (\sum_{k \in \mathbb{Z}} \|\tilde{f}[k]\|^2)^{1/2}$.

2. Problem Formulation. Consider the system in Fig. 1.1. Throughout the paper we assume that \mathcal{G} is a causal finite-dimensional LTI system given in terms of its *minimal* state-space realization

$$G(s) = \begin{bmatrix} G_v(s) \\ G_y(s) \end{bmatrix} = \left[\begin{array}{c|c} A & B \\ \hline C_v & D_v \\ C_y & 0 \end{array} \right] \quad (2.1)$$

and the estimator $\mathcal{F} : \bar{y} \mapsto u$ is shift invariant and l -causal, i.e., is in the form

$$u(t) = \sum_{i=-\infty}^{\lfloor t/h \rfloor + l} \phi(t - ih) \bar{y}[i], \quad t \in \mathbb{R} \quad (2.2)$$

for some *hold function* (interpolation kernel) $\phi(t)$ and sampling period $h > 0$. We say that \mathcal{F} is stable if it is bounded as an operator $\ell^2(\mathbb{Z}) \rightarrow L^2(\mathbb{R})$ and stabilizing if the error system \mathcal{G}_e in (1.1) is bounded as an operator $L^2(\mathbb{R}) \times \ell^2(\mathbb{Z}) \rightarrow L^2(\mathbb{R})$. The induced norm of the error system is referred to as the L^∞ norm (see [8]) and denoted as $\|\mathcal{G}_e\|_\infty$. We also assume that the realization in (2.1) satisfies

\mathcal{A}_1 : (C_y, e^{Ah}) is detectable,

\mathcal{A}_2 : $\begin{bmatrix} C_y & \Sigma \end{bmatrix}$ has full row rank.

Assumption \mathcal{A}_1 is necessary and sufficient for the existence of a stabilizing \mathcal{F} . \mathcal{A}_2 says that the measurements are not redundant and hence can be made without loss of generality. In addition, we effectively assume that $G_y(s)$ is strictly proper, which guarantees the boundedness of the ideal sampling operation.

The problem studied in this paper is formulated as follows:

RP $_{\gamma,l}$: Let signal generators \mathcal{G} and $\Sigma \geq 0$, satisfying $\mathcal{A}_{1,2}$, and a constant $l \in \mathbb{N}$ be given and let \mathcal{S} be the ideal sampler. Find whether there is a stable and stabilizing l -causal estimator \mathcal{F} of form (2.2) such that

$$\|\mathcal{G}_e\|_\infty < \gamma$$

for a given $\gamma > 0$.

RP $_{\gamma,0}$ corresponds to the filtering problem solved in [15, 20], whereas **RP $_{\gamma,\infty}$** —to the fixed-interval smoothing problem solved in [9, Sec. III].

3. Main Result. To solve the smoothing problem for this system we need two (symplectic) matrix exponentials:

$$\Lambda(t) = \begin{bmatrix} \Lambda_{11}(t) & \Lambda_{12}(t) \\ 0 & \Lambda_{22}(t) \end{bmatrix} := \exp \left(\begin{bmatrix} A & BB' \\ 0 & -A' \end{bmatrix} t \right)$$

and

$$\begin{bmatrix} \Gamma_{11}(t) & \Gamma_{12}(t) \\ \Gamma_{21}(t) & \Gamma_{22}(t) \end{bmatrix} := \exp(H_\gamma t),$$

where

$$H_\gamma := \begin{bmatrix} A & BB' \\ 0 & -A' \end{bmatrix} + \begin{bmatrix} BD'_v \\ -C'_v \end{bmatrix} (\gamma^2 I - D_v D'_v)^{-1} \begin{bmatrix} C_v & D_v B' \end{bmatrix}.$$

To shorten the notation, we omit the argument when $t = h$, so that A_{ij} and Γ_{ij} stand for $A_{ij}(h)$ and $\Gamma_{ij}(h)$, respectively.

In the solution we need two discrete algebraic Riccati equations (DAREs). The first one is the DARE associated with the Kalman filter solution:

$$Y = \Lambda_{11}(Y - YC'_y(\Sigma + C_yYC'_y)^{-1}C_yY)A'_{11} + \Lambda_{12}A'_{11}. \quad (3.1)$$

It is known [10] that if $\mathcal{A}_{1,2}$ hold, (3.1) admits a stabilizing solution $Y = Y' > 0$ for which

$$\bar{A}_1 := \Lambda_{11}(I - YC'_y(\Sigma + C_yYC'_y)^{-1}C_y) \quad (3.2)$$

is Schur. The discrete Lyapunov equation

$$X = \bar{A}'_1 X \bar{A}_1 + C'_y(\Sigma + C_yYC'_y)^{-1}C_y \quad (3.3)$$

is then always solvable by an $X = X' \geq 0$. Denote then $P := I - YC'_y(\Sigma + C_yYC'_y)^{-1}C_y$ and define the matrix

$$\begin{aligned} \begin{bmatrix} S_{\gamma,11} & S_{\gamma,12} \\ S'_{\gamma,12} & S_{\gamma,22} \end{bmatrix} &:= - \begin{bmatrix} 0 & 0 \\ 0 & \bar{A}'_1 X \bar{A}_1 \end{bmatrix} \\ &+ \begin{bmatrix} I & 0 \\ 0 & P' \end{bmatrix} \begin{bmatrix} -X & \Gamma_{22} + \Gamma_{21}PY \\ I - YX & \Gamma_{12} + \Gamma_{11}PY \end{bmatrix}^{-1} \begin{bmatrix} I & \Gamma_{21}P \\ Y & \Gamma_{11}P \end{bmatrix} \end{aligned} \quad (3.4)$$

($S_{\gamma,11} = S'_{\gamma,11} \leq 0$ and $S_{\gamma,22} = S'_{\gamma,22} \leq 0$). The second DARE,

$$Y_\gamma = S_{\gamma,12}(I + Y_\gamma S_{\gamma,22})^{-1}Y_\gamma S'_{\gamma,12} - S_{\gamma,11}, \quad (3.5)$$

is γ -dependent and its solution, which exists if γ is sufficiently large, is said to be stabilizing if $\det(I + Y_\gamma S_{\gamma,22}) \neq 0$ and the matrix $S_{\gamma,12}(I + Y_\gamma S_{\gamma,22})^{-1}$ is Schur.

The main result of this paper is then formulated as follows:

THEOREM 3.1. *Let the signal generator \mathcal{G} be given by (2.1) and assumptions $\mathcal{A}_{1,2}$ hold.*

Then $\mathbf{RP}_{\gamma,l}$ is solvable iff γ satisfies the following conditions:

1. $\gamma > \|D_v\|$,
2. $\Gamma_{12}(t) + \Gamma_{11}(t)PY$ is nonsingular $\forall t \in (0, h]$,
3. $\rho((I - Y(\Gamma_{22} + \Gamma_{21}PY)(\Gamma_{12} + \Gamma_{11}PY)^{-1})(I - YX)) < 1$,
4. *there is a stabilizing solution $Y_\gamma = Y'_\gamma \geq 0$ to the DARE (3.5) and $\rho(Y_\gamma S_{\gamma,22}) < 1$,*
5. $\rho(Y_\gamma \bar{A}'_1 X \bar{A}_1) < 1$, where $\bar{A}_i := \bar{A}_i^i$

(the first two conditions guarantee the well-posedness of (3.4)). If these conditions hold, then

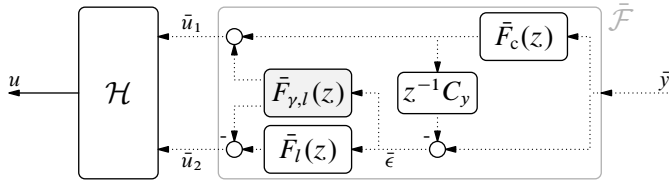


FIG. 3.1. γ -suboptimal solution

the estimator depicted in Fig. 3.1 solves the problem. It is the cascade of a discrete estimator, $\bar{\mathcal{F}}$, and a generalized zero-order hold, \mathcal{H} , with the hold function

$$\phi_h(t) := \begin{bmatrix} C_v & 0 \end{bmatrix} \Lambda(t-h) \begin{bmatrix} I & Y \\ 0 & I \end{bmatrix} \mathbb{1}_{[0,h)}(t). \quad (3.6)$$

The components of the discrete filter are

$$\bar{F}_c(z) = z \left[\begin{array}{c|c} \bar{A}_1 & A_{11} Y C_y' (\Sigma + C_y Y C_y')^{-1} \\ \hline I & 0 \end{array} \right], \quad (3.7a)$$

$$\bar{F}_{\gamma,l}(z) = z^{l+1} \left[\begin{array}{c|c} \bar{A}_1 \Delta_{l+1}^{-1} \Delta_l & \bar{A}_1 \Delta_{l+1}^{-1} Y \bar{A}_l' C_y' (\Sigma + C_y Y C_y')^{-1} \\ \hline X - \bar{A}_l' X \bar{A}_l & 0 \end{array} \right], \quad (3.7b)$$

$$\bar{F}_l(z) = \sum_{i=0}^{l-1} \bar{A}_{l-1-i}' C_y' (\Sigma + C_y Y C_y')^{-1} z^{l-i}, \quad (3.7c)$$

where $\Delta_i := I - Y \bar{A}_i' X \bar{A}_i$.

Proof. Omitted because of space limitations. \square

Some remarks are in order:

Remark 3.1 (solvability conditions). The first four conditions of Theorem 3.1 do not depend on the smoothing lag l . These are the necessary and sufficient conditions for the solvability of the L^∞ fixed-interval smoothing problem ($l \rightarrow \infty$). The fifth solvability condition of Theorem 3.1 reflects then constraints imposed by a finite preview. Because \bar{A}_1 is Schur, $\rho(Y \bar{A}_l' X \bar{A}_l)$, as a function of l , is upperbounded by an exponentially decreasing function. Hence, whenever Y_γ is bounded, there exist a finite l for which the causality constraint becomes inactive. ∇

Remark 3.2 (solvability for $l = \infty$). It can be shown [9] that γ satisfies the first four conditions of Theorem 3.1 iff

$$\gamma > \gamma_h := \left\| \begin{bmatrix} \mathcal{G}_v & 0 \end{bmatrix} - \mathcal{G}_v (\mathcal{S}\mathcal{G}_y)^* (\Sigma + \mathcal{S}\mathcal{G}_y (\mathcal{S}\mathcal{G}_y)^*)^{-1} \begin{bmatrix} \mathcal{S}\mathcal{G}_y & \Sigma^{1/2} \end{bmatrix} \right\|_{L^2 \times \ell^2 \rightarrow L^2}.$$

In the case when $\Sigma > 0$, this γ_h can be characterized via the self-adjoint operator $\check{O}_\gamma(e^{j\theta})$, described by the following two-point boundary condition system [5]:

$$\begin{cases} \dot{x}(t) = \begin{bmatrix} A & BB' \\ 0 & -A' \end{bmatrix} x(t) + \begin{bmatrix} BD_v' \\ -C_v' \end{bmatrix} u(t), & e^{j\theta} x(0) = \begin{bmatrix} I & 0 \\ C_y' \Sigma^{-1} C_y & I \end{bmatrix} x(h) \\ y(t) = \begin{bmatrix} C_v & D_v B' \end{bmatrix} x(t) + D_v D_v' u(t) \end{cases}$$

Namely, $\gamma > \gamma_h$ iff $\check{O}_\gamma(e^{j\theta}) < \gamma^2 I$ for all $-\pi < \theta \leq \pi$. Thus, γ_h is the largest γ for which the symplectic matrix

$$M_\gamma := \begin{bmatrix} I & 0 \\ C_y' \Sigma^{-1} C_y & I \end{bmatrix} \begin{bmatrix} \Gamma_{11} & \Gamma_{12} \\ \Gamma_{21} & \Gamma_{22} \end{bmatrix}$$

has unit circle eigenvalues. The matrix M_γ is actually similar to the symplectic matrix associated with the sampled-data H^∞ filtering Riccati equation in [15, 20] and it becomes the symplectic matrix associated with (3.1) as $\gamma \rightarrow \infty$. ∇

Remark 3.3 (recovering the L^2 solution). The only difference between the the L^∞ estimator of Theorem 3.1 and the L^2 solution of [10] is the presence of $\bar{F}_{\gamma,l}$, the gray block in Fig. 3.1, in the former. This block vanishes in two limiting cases. First, because \bar{A}_1 is Schur, $\lim_{l \rightarrow \infty} \bar{A}_l = 0$ and the fixed-interval solution is independent of γ (provided it satisfies the first four conditions of Theorem 3.1, of course) and approaches the L^2 -optimal solution. Second, it follows from the proof of Theorem 3.1 that

$$\lim_{\gamma \rightarrow \infty} \begin{bmatrix} S_{\gamma,11} & S_{\gamma,12} \\ S_{\gamma,12}' & S_{\gamma,22} \end{bmatrix} = \begin{bmatrix} 0 & \bar{A}_1 \\ \bar{A}_1' & 0 \end{bmatrix}.$$

In this case (3.5) reads $Y_\gamma = \bar{A}_1 Y \bar{A}_1'$ and its stabilizing solution is $Y_\gamma = 0$. Hence, the gray block vanishes for $\gamma \rightarrow \infty$ too (in this case the conditions of Theorem 3.1 hold $\forall h > 0$). ∇

4. Example: causal L^∞ cubic splines. Consider the problem with

$$G_v(s) = G_y(s) = \frac{1}{s^2} = \left[\begin{array}{cc|c} 0 & 1 & 0 \\ 0 & 0 & 1 \\ \hline 1 & 0 & 0 \end{array} \right] \quad \text{and} \quad \Sigma = 0,$$

which does satisfy assumptions $\mathcal{A}_{1,2}$. Without loss of generality we may assume that $h = 1$. In the non-causal case ($l = \infty$) this setting reproduces the *cardinal cubic B-splines* [22], which are perhaps the most thoroughly studied polynomial splines. It is worth emphasizing that in that case the L^2 and L^∞ criteria result in identical estimators, which is a known property of non-causal solutions [6, §10.4.2]. Then, in [10], we studied the L^2 version of the problem under causality constraints, i.e., in the fixed-lag smoothing setup. The impulse response of the resulting estimators could then be regarded as causal cubic splines. If causality constraints are present, L^2 (mean-square) solutions are no longer identical to L^∞ (minmax) solutions. It is therefore of interest to see how cardinal cubic splines evolve under causality constraints in the L^∞ setting. This is the main goal of this section.

Although $\mathbf{RP}_{\gamma,l}$ studies a suboptimal solution ($\|\mathcal{G}_e\|_\infty < \gamma$), in this section we consider the optimal case corresponding to $\|\mathcal{G}_e\|_\infty \leq \gamma$. This is done by addressing the limiting γ , in which case the DARE (3.5) no longer has a stabilizing solution, but still has a real positive definite one. In general, this might be a delicate procedure [16], but it works for this specific example painlessly, with the last condition of Theorem 3.1 replaced with $\rho(Y_\gamma \bar{A}_1' X \bar{A}_1) \leq 1$.

First, let us calculate the matrices associated with the L^2 solution. They are

$$\Lambda(t) = \left[\begin{array}{cc|cc} 1 & t & -t^3/6 & t^2/2 \\ 0 & 1 & -t^2/2 & t \\ \hline 0 & 0 & 1 & 0 \\ 0 & 0 & -t & 1 \end{array} \right], \quad Y = \frac{1}{6} \begin{bmatrix} 2 + \sqrt{3} & 3 + \sqrt{3} \\ 3 + \sqrt{3} & 6 + \sqrt{3} \end{bmatrix}, \quad \bar{A}_1 = \begin{bmatrix} \sqrt{3} - 3 & 1 \\ \sqrt{3} - 3 & 1 \end{bmatrix},$$

$$P = \begin{bmatrix} 0 & 0 \\ \sqrt{3} - 3 & 1 \end{bmatrix}, \quad \text{and} \quad X = \begin{bmatrix} 6\sqrt{3} - 6 & 3 - 3\sqrt{3} \\ 3 - 3\sqrt{3} & \sqrt{3} \end{bmatrix}.$$

Then the hold function defined by (3.6) is

$$\phi_h(t) = [1 \quad -1 + t \quad t(-t^2 + 3t + \sqrt{3})/6 \quad t(3t + \sqrt{3})/6] \mathbb{1}_{[0,1)}(t).$$

Using the arguments of Remark 3.2, it can be shown that the minimal achievable $\|\mathcal{G}_e\|_\infty$ in the non-causal case is $\gamma = 1/\pi^2 \approx 0.1013$. For this γ the first three conditions of Theorem 3.1 hold, the matrix

$$\Gamma = \left[\begin{array}{cc|cc} (\sinh \frac{\pi}{2})^2 & \frac{1}{2\pi} \sinh \pi & -\frac{1}{3\pi^3} \sinh \pi & \frac{1}{2\pi^2} (1 + \cosh \pi) \\ \frac{\pi}{2} \sinh \pi & (\sinh \pi)^2 & -\frac{1}{2\pi^2} (1 + \cosh \pi) & \frac{1}{2\pi} \sinh \pi \\ \hline -\frac{\pi^3}{2} \sinh \pi & -\frac{\pi^2}{2} (1 + \cosh \pi) & (\sinh \frac{\pi}{2})^2 & -\frac{\pi}{2} \sinh \pi \\ \frac{\pi^2}{2} (1 + \cosh \pi) & \frac{\pi}{2} \sinh \pi & -\frac{1}{2\pi} \sinh \pi & (\sinh \frac{\pi}{2})^2 \end{array} \right]$$

and then

$$S_\gamma = \left[\begin{array}{cc|cc} -15.410 & -17.939 & -4.271 & 3.369 \\ -17.939 & -20.935 & -4.624 & 3.647 \\ \hline -4.271 & -4.624 & -1.045 & 0.825 \\ 3.369 & 3.647 & 0.825 & -0.650 \end{array} \right] \quad \text{and} \quad Y_\gamma = \begin{bmatrix} 25.900 & 29.296 \\ 29.296 & 33.229 \end{bmatrix}.$$

The latter is positive definite and such that the eigenvalues of $S_{\gamma,12}(I + Y_\gamma S_{\gamma,22})^{-1}$ are $\{-1, 0\}$ ($S_{\gamma,12}$ is singular and Y_γ is a semi-stabilizing solution of (3.5) now).

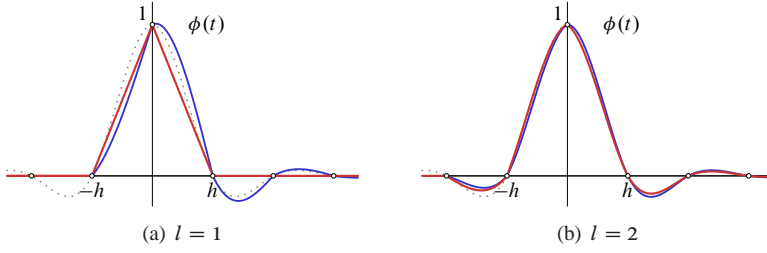


FIG. 4.1. Hold functions $\phi(t)$ (red lines: L^∞ , blue lines: L^2 , dotted gray lines: $l = \infty$)

Now,

$$\rho(Y_\gamma \bar{A}_1' X \bar{A}_1) = \rho \left(\begin{bmatrix} \pi^2(3 + \sqrt{3})/6 & -\pi^2(2 + \sqrt{3})/6 \\ \pi^2 - 3 + \sqrt{3} & 1 - \pi^2(3 + \sqrt{3})/6 \end{bmatrix} \right) = 1,$$

which implies that the fixed-interval performance $\gamma = 1/\pi^2$ is achievable for every $l \in \mathbb{N}$.

We consider two cases: $l = 1$ and $l = 2$. The discrete filters \bar{F} in Fig. 3.1 for these smoothing lags have transfer functions

$$\bar{F}(z) = \begin{bmatrix} z \\ z-1 \\ 0 \\ 0 \end{bmatrix} \quad \text{and} \quad \bar{F}(z) = \begin{bmatrix} -z^3 + (4 - \sqrt{3})z^2 + 2(2 + \sqrt{3})z - 2 - \sqrt{3} \\ -(z-1)(z^2 - (3 - \sqrt{3})z - 3 - \sqrt{3}) \\ -6(z-1)^3 \\ 6z(z-1)^2 \end{bmatrix} \frac{1}{4z+1},$$

respectively. Note that the dynamics of the causal part of \bar{F} depend on l . In fact, as l increases, their pole approaches $\alpha := \sqrt{3} - 2$ via the sequence $\{-\frac{1}{4}, -\frac{4}{15}, -\frac{15}{56}, -\frac{56}{209}, -\frac{209}{780}, \dots\}$. This is in contrast to the L^2 case, where the causal pole is located at α at every l .

The resulting hold functions are presented in Fig. 4.1 by red lines. For the sake of comparison, blue lines there show the corresponding L^2 solutions of [10] and dotted lines show the fixed-interval solution (cardinal cubic B-spline). It is seen from Fig. 4.1(a) that in the case of $l = 1$ we end up with the *predictive first-order hold* (linear interpolator), whose hold function

$$\phi(t) = (1 - |t|)\mathbb{1}_{[-1,1]}(t)$$

is linear in t . This is surprising because this function is both L^2 and L^∞ optimal also in the case when $G_v(s) = G_y(s) = 1/s$ for every $l \in \mathbb{N}$ [11, Sec. III]. For $l > 1$ the optimal holds of Theorem 3.1 are cubic in t and are qualitatively closer to the corresponding L^2 solutions.

It is worth emphasizing that the L^∞ hold functions shown in Fig. 4.1, as well as every L^∞ hold for $l > 2$, attain the very same $\|\mathcal{G}_e\|_\infty = 1/\pi^2$. Yet as l increases, the L^2 performance improves, see [13, Sec. 4]. For example, if $l = 1$, the L^∞ estimator attains $\|\mathcal{G}_e\|_2 \approx 0.1054$, which amounts to some 120% of the optimal L^2 performance level for $l = 1$. If $l = 2$, we have $\|\mathcal{G}_e\|_2 \approx 0.0773$, which amounts to about 101.3% of the corresponding optimal value.

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